Digital Automatic Control System with PID Controller

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Introduction

Mass production of microprocessors reduced their cost. Because of this reason they found the new areas of application: to fulfill various computations, to solve control and signal processing task. Nowadays they can be found not just in industry but also in transportation systems and many more complex domestic appliances.

The control problems can be solved by application of analogue means, but digital controllers have some advantages [1]: their parameters remains stable for long time; realized function can be changed only by changing program; digital circuits are robust for noise; single controller in accordance with the speediness, may be capable to control severe plants.

Industrial digital controllers usually are realized on the base of programmable logic controllers [2]. An in advance preprogrammed transfer function of many popular controllers is the main advantage of those. Another advantage of programmable logic controllers is possibility to deal with various types of digital and analog input and output signals. In the mobile and domestic appliances as well as stand-alone devices they are used seldom due to large size and high power consumption. Therefore this area at the present is covered by microcontrollers.

Microcontrollers are used in automatically controlled products and devices, such as automobile engine control systems, implantable medical devices, remote controls, office machines, power tools and toys.

A microcontroller can be consider as a self-contained system with a processor, memory and peripherals and be used as embedded system. Microcontrollers use smaller bit words, but their integrated circuit contains a processor core, programmable input/output peripherals and also other functional units, as memory, timers, USB, UART and other interfaces, PWM signal generator and analog-digital converters. Due to small size and small power consumption they are well-suited for domestic, mobile and long lasting battery applications. The most popular are eight-bit word microcontrollers. Usually they are applied for simple task, where floating point operations are not required. For more complicated tasks 16 or 32 bit microcontrollers are applied [3].

Proportional-integral-derivative (PID) controller is the base of universal controllers. It is applied in many industrial control systems and has a lot of forms depending on application [4, 5], and for its tuning many laws are used [6].

The article deals with the problems appearing by programming of PID controller in eight-bit word microcontroller and ways of their solving.

PID algorithm

PID controller usually is used in closed loop systems and forms system control signal $y$ according to error $e$ of controlled parameter $i$. Block diagram of closed loop system with PID controller is shown in Fig. 1.

![PID controller block diagram](image)

Fig. 1. Closed loop system with PID controller

Relationship between input and output on continuous time classical PID controller is described as [1]

$$y = K_p \left( e + \frac{1}{T_i} \int e \, dt + T_d \frac{de}{dt} \right),$$

(1)

where $K_p$, $T_i$ and $T_d$ are positive parameters, which are respectively referred to as proportional gain, integral time and derivative time.

Relationship (1) should be replaced by second order difference equation. It takes the form as

$$y_n = -a_1 y_{n-1} - a_2 y_{n-2} + b_0 e_n + b_1 e_{n-1} + b_2 e_{n-2};$$

(2)

where $a_1$, $a_2$, $b_0$, $b_1$, $b_2$ are constants, variables $y_n$, $y_{n-1}$, $y_{n-2}$, $e_n$, $e_{n-1}$ and $e_{n-2}$ correspond to controller output and input signals at the time instant $nT, (n-1)T$ ir $(n-2)T$, where $T$ is sampling period and $n \in N$. 

The PID controller deals with sampled input signal $e_n$ which is discontinuous function. Differentiating of such kind function may causes large errors. In order to avoid this disadvantage the ideal derivative part of controller is recommended to replace by real that [1]. In this way the function realized by controller is described as

$$y(s) = K_p \left( 1 + \frac{1}{T_n s} + \frac{T_T s}{T_T s + 1} \right) e(s);$$

(3)

where $T_t$ is time constant of real differentiation block, $y(s)$ and $e(s)$ is respectively the output and input signals in frequency domain.

The transfer function of PID controller is

$$W_{PID}(s) = \frac{y(s)}{e(s)} = K_p \left( 1 + \frac{1}{T_n s} + \frac{T_T s}{T_T s + 1} \right).$$

(4)

Chosen function is realized by controller. Coefficients of and its difference equation are calculated from expressions [1]:

$$h_0 = \frac{K_p}{1 + T_n / T} \left( 1 + \frac{T + T_n}{2 T_n} + \frac{T_v + T_t}{T} \right);$$

$$h_1 = \frac{K_p}{1 + T_n / T} \left( -1 + \frac{T}{2 T_n} - 2 \frac{T_v + T_t}{T} \right);$$

$$b_2 = \frac{K_p}{1 + T_n / T} \left( \frac{T_T s}{2 T_n} + \frac{T_v + T_t}{T} \right);$$

$$a_1 = -\frac{T_t}{T_T s};$$

$$a_2 = \frac{T_T}{T_T s + 1}.$$  

(5)

Analysis of difference equation (2) and coefficients expressions (5) shows that realization of PID controller requires multiplication and adding of real numbers.

**Design of the controller of mechatronic system**

Controller is devoted for magnetic bearings system where the current control in the field coil is required. The plant is described by first order transfer function

$$W(s) = \frac{k}{T_s s + 1}.$$  

(6)

where $k$ is gain and $T$ time constant.

The transfer function of closed loop system, shown in Fig. 1 is

$$W_s(s) = K_p \frac{L \mu s^n + T_T s + 1}{T_n \left[ L + K_p \mu T_T s^n + T_n \left[ R + K_p \mu s^n + K_p \right] \right]}.$$  

(7)

The plant is the coil of electromagnet with time constant $T=L/R$ and gain $k=1/R$, where $L$ and $R$ is coil inductance and resistance respectively. Assumed parameters are: $L=0.047$ H and $R=15$ Ω. The current settling time will be approximately $4T/R=12$ ms. If desirable step response settling time should last no longer 6 ms, and overshoot should be not greater than 10%, then on the base of classical control theory the desired coefficients of close loop system can be calculated in this way:

$$\begin{align*}
T_n \left( \frac{L}{K_p} + T_v \right) &\approx 1.96 \cdot 10^{-6}; \\
T_n \left( \frac{R}{K_p} + 1 \right) &\approx 0.02.
\end{align*}$$  

(8)

Set of equation (8) has 3 variables, therefore it cannot be solved explicit. By setting $K_p=40$, we get $T_s=0.00145$ and $T_v=1.73\cdot10^{-6}$.

Step responses of current with controller and without it are presented in Fig. 2. Simulation results show the overshoot being about 9% and settling time not exceeding 6 ms.

![Fig. 2. Step responses of current](image)

All coefficients of difference equation depend on sampling time $T$. Algorithm of controller has to be programmed into microcontroller ATmega16. This microcontroller has integrated analog-digital converters whose maximal sampling frequency without losing accuracy is 15 kHz, therefore minimal sampling time is about $66,667$ μs. Coefficients (5) of difference equation for this and some other values of sampling time are presented in table 1.

<table>
<thead>
<tr>
<th>$T_s$ (μs)</th>
<th>66,667</th>
<th>70</th>
<th>100</th>
<th>500</th>
</tr>
</thead>
<tbody>
<tr>
<td>$b_0$</td>
<td>123</td>
<td>120</td>
<td>100</td>
<td>60.2</td>
</tr>
<tr>
<td>$b_1$</td>
<td>-212</td>
<td>-205</td>
<td>-162</td>
<td>-61.3</td>
</tr>
<tr>
<td>$b_2$</td>
<td>90.3</td>
<td>86.8</td>
<td>64.5</td>
<td>14.4</td>
</tr>
<tr>
<td>$a_1$</td>
<td>-1.21</td>
<td>-1.20</td>
<td>-1.15</td>
<td>-1.03</td>
</tr>
<tr>
<td>$a_2$</td>
<td>0.206</td>
<td>0.198</td>
<td>0.147</td>
<td>0.0333</td>
</tr>
<tr>
<td>$\alpha$</td>
<td>1029</td>
<td>1035</td>
<td>1104</td>
<td>1839</td>
</tr>
<tr>
<td>$log_{10}n$</td>
<td>10</td>
<td>10</td>
<td>10.1</td>
<td>10.8</td>
</tr>
</tbody>
</table>

The same table contains additional parameters, which will be valuable in estimation of the word length for storage of coefficients and results. One of the parameters is the ratio of absolute values of maximal and minimal coefficients $\alpha$. 

Table 1. Values of difference equation coefficients
Analysis of coefficients values, presented in Table 1, shows that the ratio between the maximal and the minimal values lies in the range of 1÷2 thousands. For binary representation of this range 10÷11 bit length words are required.

PID controller is intended to be implemented in Atmel company’s 8 bit microcontroller ATmega16, but it can be applied for the other megaAVR family controllers having the same command system.

As it was mentioned, 8 bit length word is insufficient to represent the coefficients of difference equation, therefore several bytes to store the numbers should be used. For storage of real numbers fixed or floating point formats can be used. The floating point representation is used by all modern computers, but it demands a lot of computation capability. Despite very limited computation capability our system should be able to work as real time system, because of that the fixed point representation of real numbers is chosen. According to the Table 1, minimal length of the word should be 10 bits. For representing them by 8 bit microcontroller 2 bytes length word was chosen. Bytes assignment for representation sign, integer and fractional parts of the number are listed in the Table 2.

Table 2. Representation of real numbers

<table>
<thead>
<tr>
<th>Bits</th>
<th>Number of bits</th>
<th>Meaning</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>1</td>
<td>Sing (+ or –)</td>
</tr>
<tr>
<td>7 - 14</td>
<td>8</td>
<td>Integer part (0 – 255)</td>
</tr>
<tr>
<td>0 - 6</td>
<td>7</td>
<td>Fraction part</td>
</tr>
</tbody>
</table>

As the limited length of the word was chosen, the coefficients of difference equation and intermediate results can be represented with limited accuracy, i.e. the rounding errors appear. For example, the coefficient \(a_2\) at \(T=66,667\) μs is equal to 0,206. In chosen system it will be represented by 0x001A hexadecimal number, which has the decimal value 0,2031. Rounding error in this case is

\[
0,206 - 0,2031 \approx 1,4\%.
\] (10)

In the same way calculated rounding errors of the other coefficients are presented in Table 3.

Table 3. Rounding errors

<table>
<thead>
<tr>
<th>(T, \mu s)</th>
<th>66,667</th>
<th>70</th>
<th>100</th>
<th>500</th>
</tr>
</thead>
<tbody>
<tr>
<td>(b_0)</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0,008</td>
</tr>
<tr>
<td>(b_1)</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0,006</td>
</tr>
<tr>
<td>(b_2)</td>
<td>0,003</td>
<td>0,004</td>
<td>0</td>
<td>0,01</td>
</tr>
<tr>
<td>(a_1)</td>
<td>3,3</td>
<td>0,5</td>
<td>0,1</td>
<td>0,6</td>
</tr>
<tr>
<td>(a_2)</td>
<td>1,4</td>
<td>0,3</td>
<td>0,6</td>
<td>6</td>
</tr>
</tbody>
</table>

Results shown in the Table 3 indicates, that maximal values of rounding errors, caused by choosing of 2 bytes length fixed point format for representing of real numbers, reach 6 %. This error could be reduced by choosing longer word length for representation of real numbers. It should be noticed that the larger error is characteristic for small magnitude coefficients.

During the development of controller it is important to choose optimal sampling frequency and numbers storage format, because in this way the program may be optimized and errors can be reduced.

Despite all given remarks in Table 2 shown number format matches all engineering calculations and allows to represent the coefficients of difference equation with suitable accuracy.

Algorithm of multiplication

ATmega16 controller, as well as others megaAVR family controllers, has only 8 bit multiplication command. Our case deals with multiplication of 16 bits (2 bytes), therefore it is convenient use the partition

\[
\Pi = (256a_1 + a_0)(256b_1 + b_0) =
\]

\[
= 65536a_1b_1 + 256(a_1b_0 + a_0b_1) + a_0b_0;
\] (8)

where \(a_0, b_0\) – low bytes of the multipliers, \(a_1, b_1\) – high bytes of the multipliers.

Algorithm of subprogram realizing multiplication

![Algorithm of subprogram realizing multiplication](image)

Algorithm of subprogram realizing multiplication is given in Fig. 2. Here \(R_x\) corresponds to microcontroller registers, \(C\) – carry flag attribute. It can be seen that the subprogram uses only 8 microcontroller registers, where each two are used for storage of multipliers and four are used for storage of product. 20 MHz clock rate microcontroller fulfills it in time less than 1 μs. Solution of difference equation (2) takes 12 μs.

Step responses of current \(i\) and error \(e\) (controller input signal) at different sampling time was obtained by simulation and presented in Fig. 3.

Step response curves in Fig. 3 indicate, that overshoot increase together with increasing sampling time. At \(T=100\) μs maximum overshoot values is 9%, but at \(T=500\) μs overshoot rise up to 12 % and exceeds allowed 10 % value.
Fig. 3. Step response of current and error of current

Step response of the controller output signal at the beginning of the transient process is presented in Fig. 4.

Fig. 4. Step response of controller output signal

Fig. 4 shows that with decreasing of sampling time $T$ the maximal value of output signal increases. In the real system the output signal always is limited, therefore reducing of sampling time is not useful for the system.

Conclusions

Analysis shows, that 8 bit microcontroller capability allows to program PID controller with sampling time no less than 100 μs. Despite availability of microcontroller ADC to operate faster, the speediness of microcontroller arithmetic logic unit is not sufficient to solve difference equation in real time.

Due to restricted resources of the microcontroller it should be programmed for each task individually. Even small change in conditions of the same task, for example change of sampling time, can result increasing of the error.

Greater sampling time gives in overshoot increasing and non continuity of step response at the beginning of transients. Therefore if the plant is sensitive for dynamic overloads, the small as possible sampling time is recommended.

Reduction of sampling time increases controller output voltage in the beginning of the step response due to derivate action of controller.

References


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The design process of digital automatic control system with PID controller is considered. The solution of problems related with implementation of PID control algorithm into general purpose 8-bit microcontroller is discussed. Simulation results demonstrating performance of system are presented. Ill. 4, bibl. 6, tabl. 3 (in English; abstracts in English and Lithuanian).


Projektuojama sistema su skaitmeniniu PID reguliatoriumi. PID reguliatorius įrengiamas bendros paskirties 8 bitų mikrokontroleryje. Aptariami sprendimai susiję su diskretizacijos periodo ir realiųjų skaičių reprezentavimo būdu mikrokontroleryje pasirinkimu. Pateiki ir aptarti uždaros sistemos skaitinio imitavimo rezultatai. Il. 4, bibl. 6, lent. 3 (anglų kalba; santraukos anglų ir lietuvių k.).